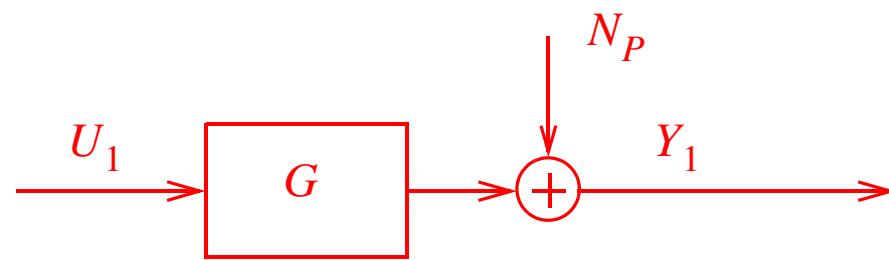


Frequency Domain Maximum likelihood Estimation of Linear Dynamic Errors-in-Variables Models

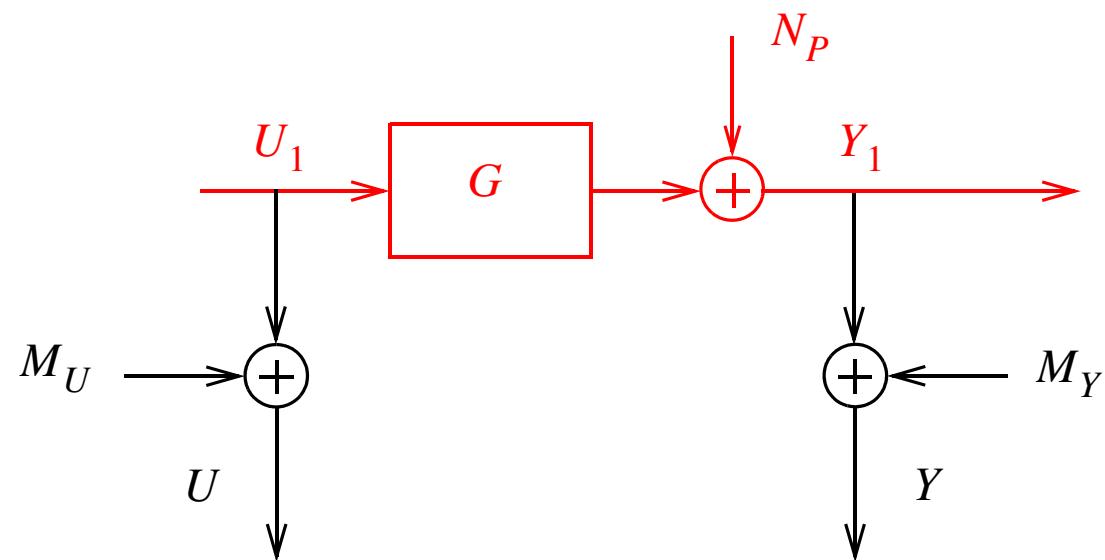
R. Pintelon, and J. Schoukens

Vrije Universiteit Brussel, dept. ELEC

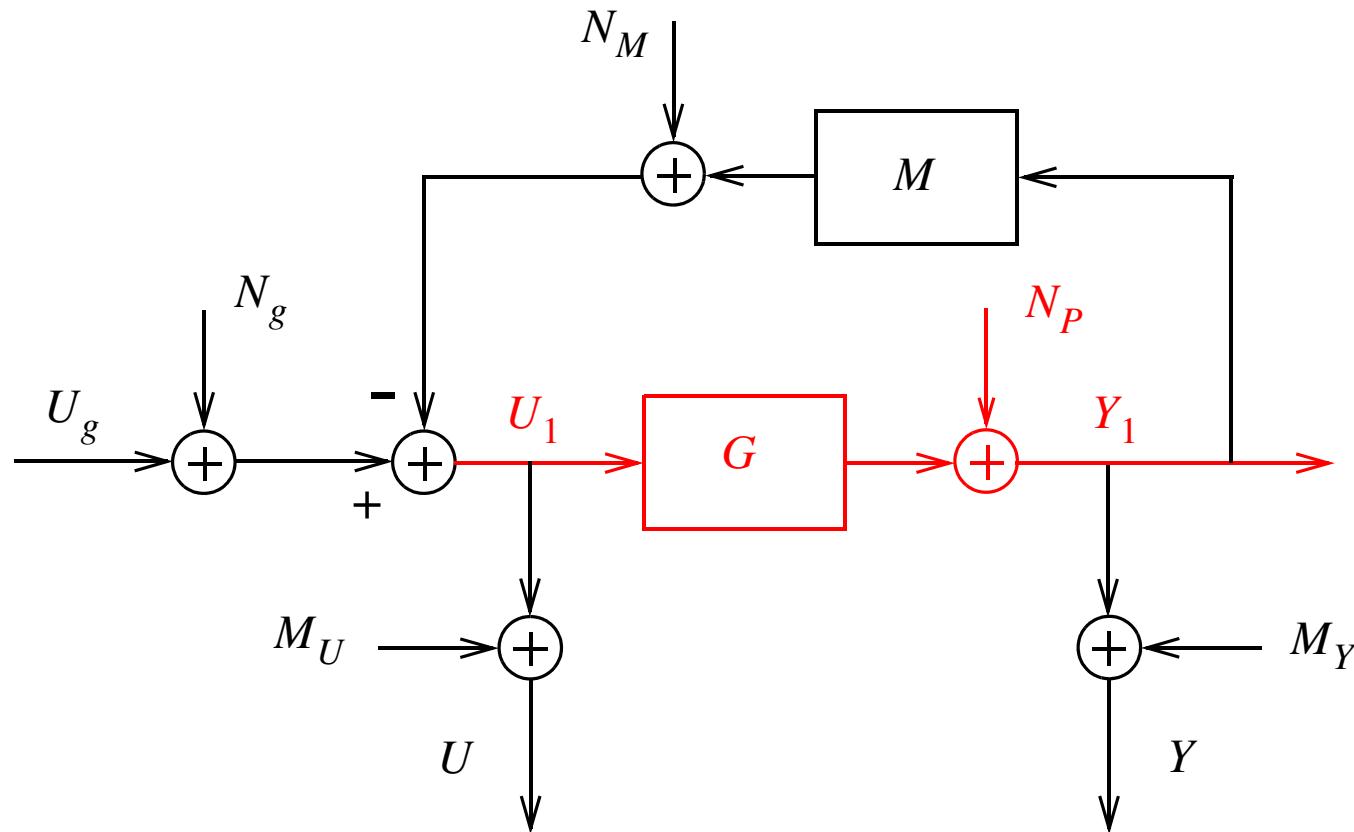
Open Loop Generalised-Output-Error Problem



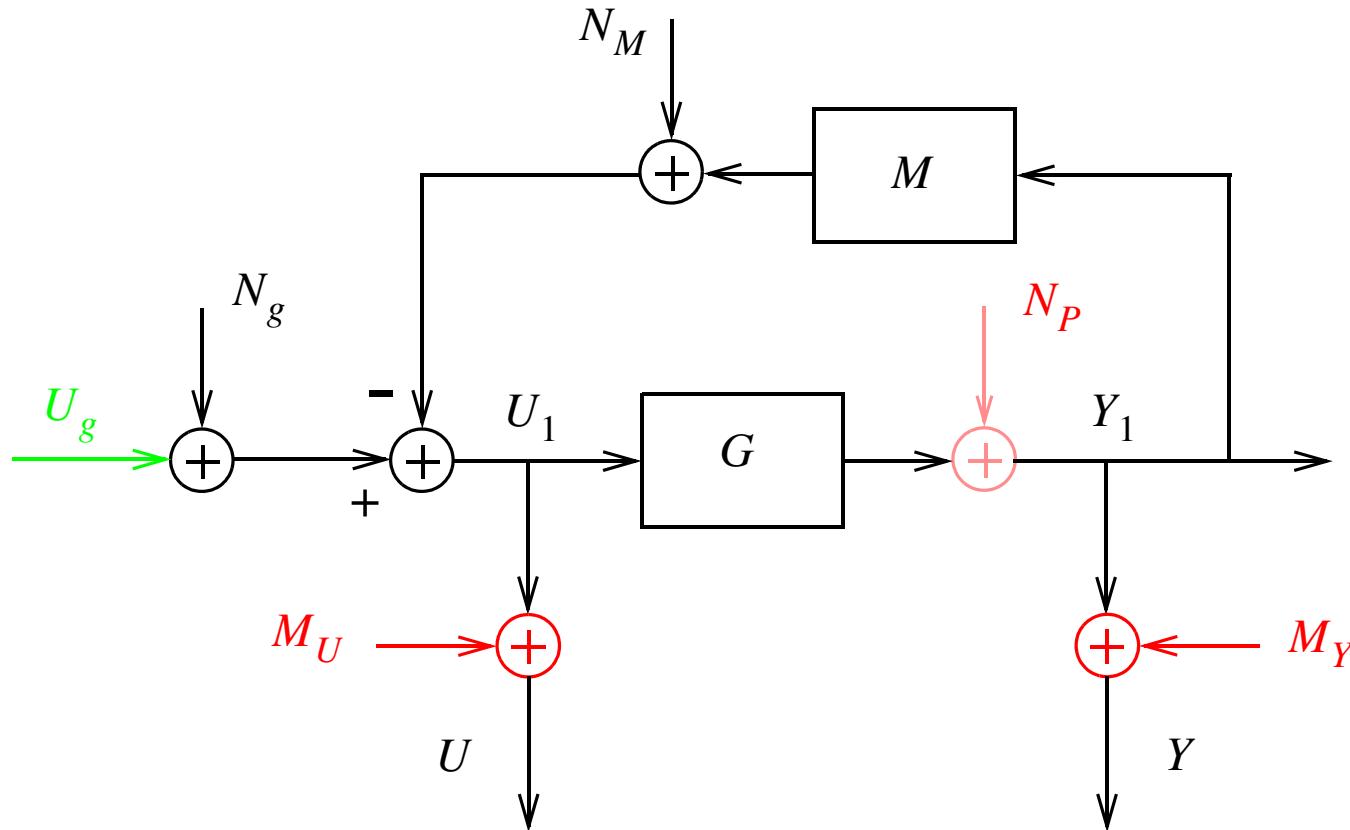
Open Loop Errors-in-Variables Problem



Closed Loop Errors-in-Variables Problem



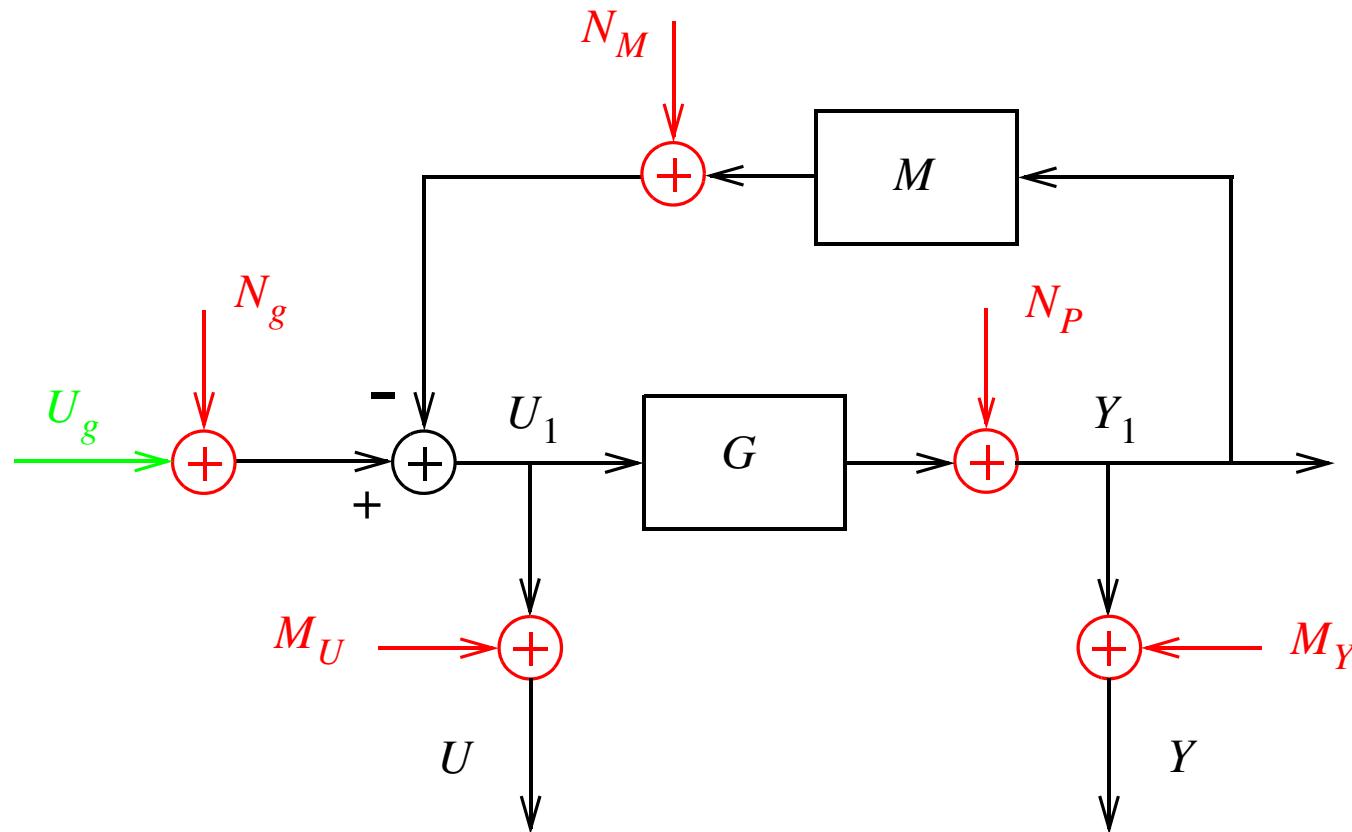
Closed Loop Errors-in-Variables Problem



Arbitrary excitation

$$\begin{cases} U_0 = \frac{1}{1+GM}(U_g + N_g - N_M - MN_P) \\ Y_0 = \frac{G}{1+GM}(U_g + N_g - N_M - MN_P) \end{cases} \quad \text{and} \quad \begin{cases} N_U = M_U \\ N_Y = M_Y + \frac{1}{1+GM}N_P \end{cases}$$

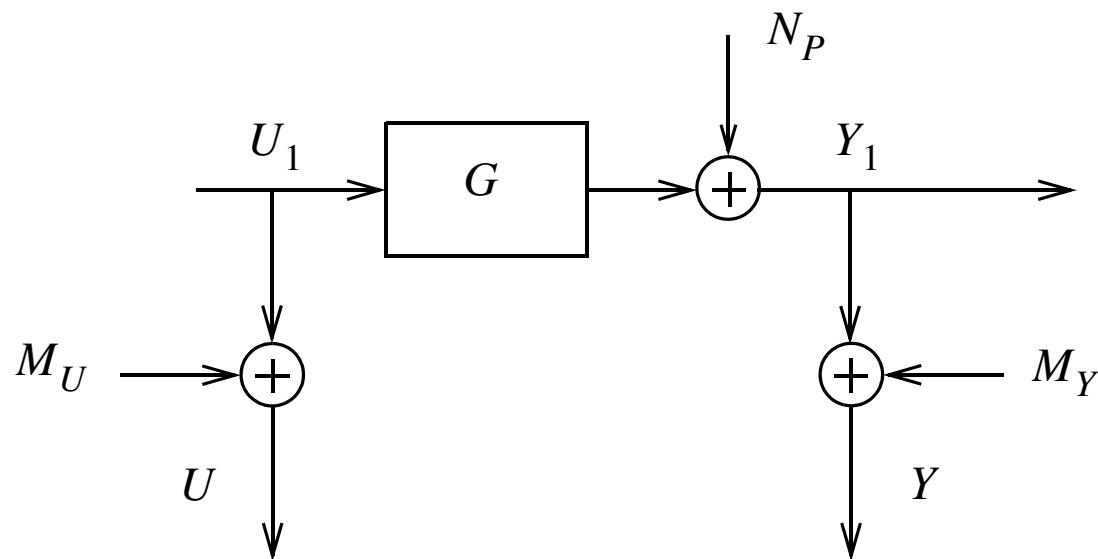
Closed Loop Errors-in-Variables Problem



Periodic excitation

$$\begin{cases} U_0 = \frac{1}{1+GM} U_g \\ Y_0 = \frac{G}{1+GM} U_g \end{cases} \text{ and } \begin{cases} N_U = M_U + \frac{1}{1+GM}(N_g - N_M) - \frac{M}{1+GM}N_P \\ N_Y = M_Y + \frac{G}{1+GM}(N_g - N_M) + \frac{1}{1+GM}N_P \end{cases}$$

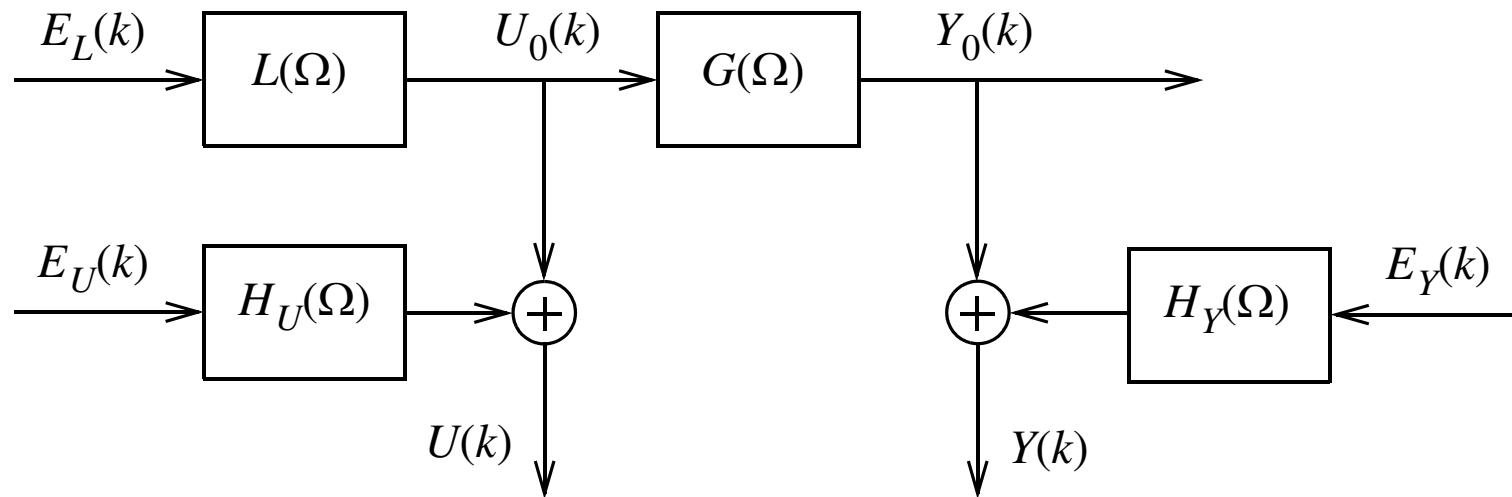
This Paper: Simplified Errors-in-Variables Problem



Assumptions

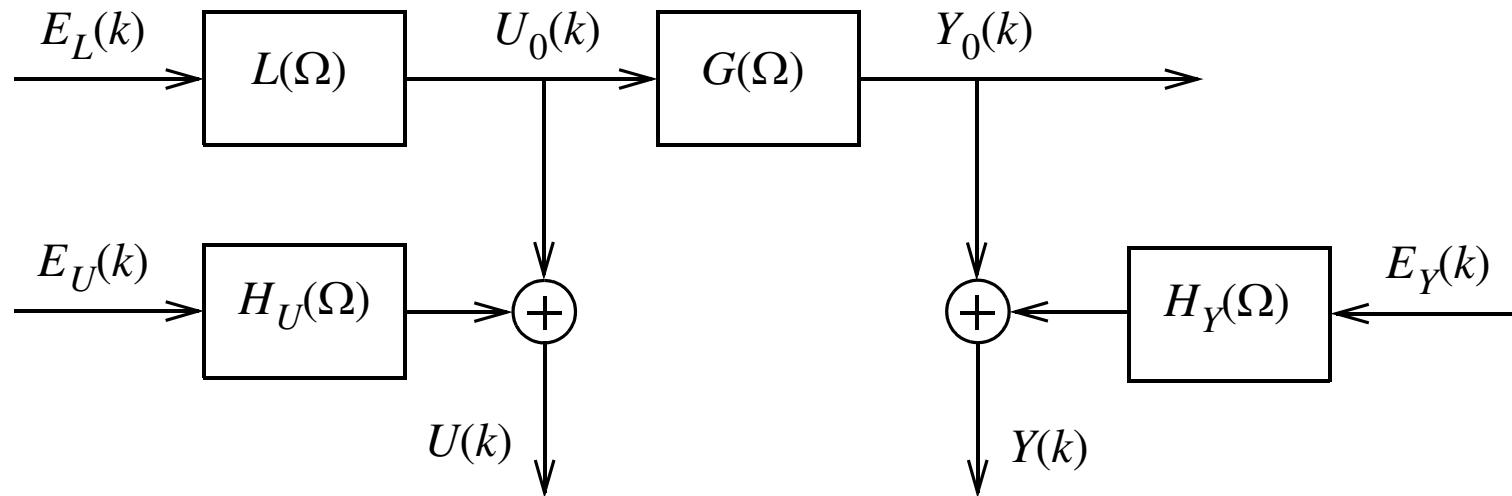
- Open loop
- Filtered white noise excitation
- Independent, filtered white noise input/output errors

Errors-in-Variables Framework



$$\Omega = \begin{cases} z^{-1} & \text{DT} \\ s & \text{CT} \end{cases}$$

Errors-in-Variables Framework

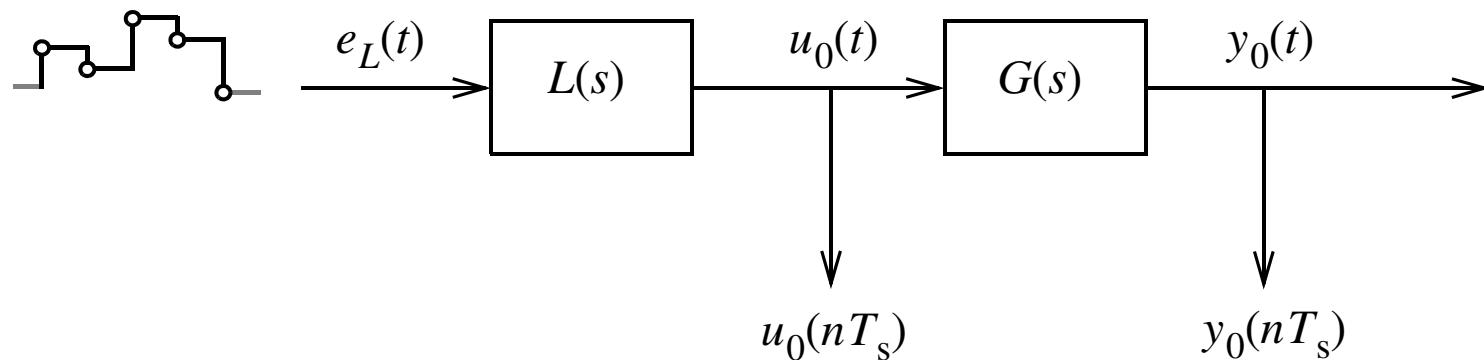


$$\Omega = \begin{cases} z^{-1} & \text{DT} \\ s & \text{CT} \end{cases}$$

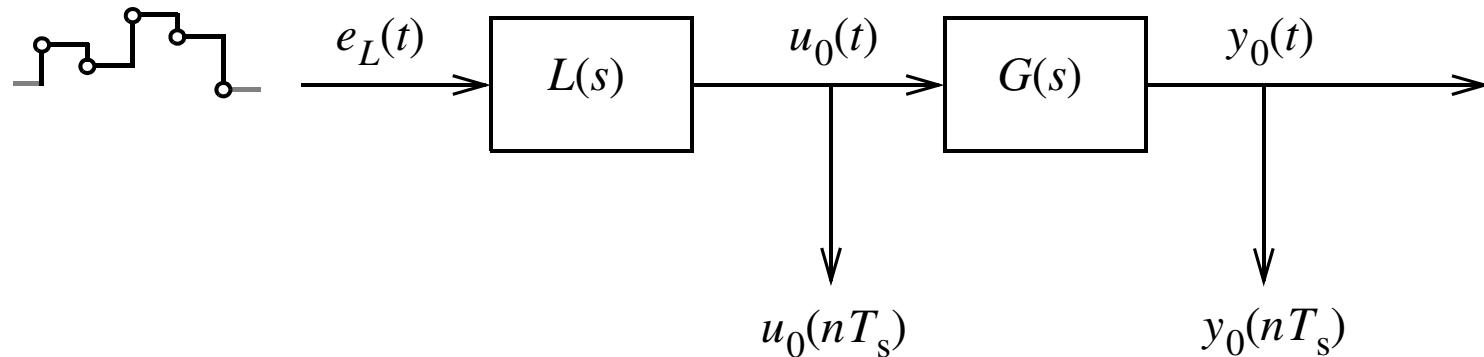
with $E_U(k)$, $E_Y(k)$, and $E_L(k)$

- Mutually independent
- Independent over k
- Circular complex, normally distributed

Errors-in-Variables Framework

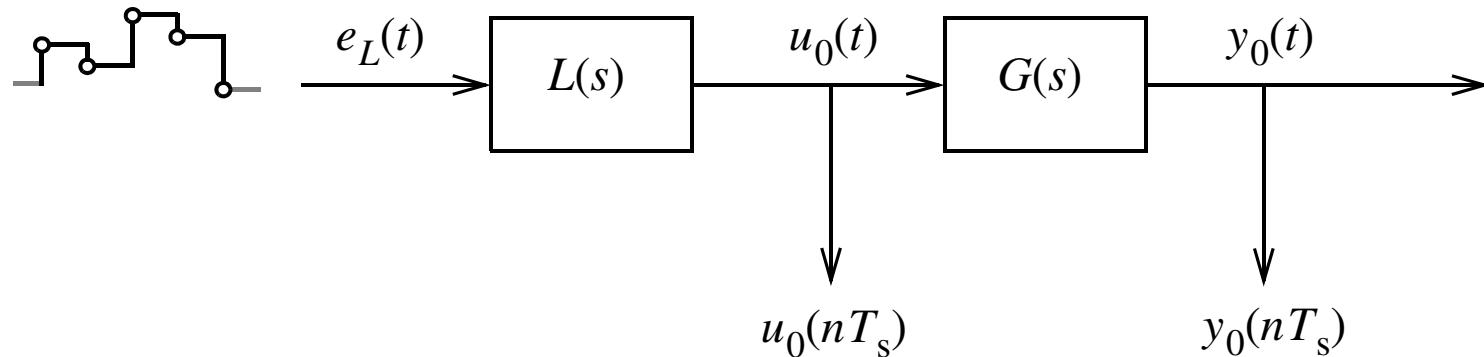


Errors-in-Variables Framework



$$G_1(z^{-1}) = \frac{Z\{y_0(nT_s)\}}{Z\{u_0(nT_s)\}} = \frac{\frac{Z\{y_0(nT_s)\}}{Z\{e_L(nT_s)\}}}{\frac{Z\{u_0(nT_s)\}}{Z\{e_L(nT_s)\}}} = \frac{(1 - z^{-1})Z\{L^{-1}\{L(s)G(s)/s\}\}}{(1 - z^{-1})Z\{L^{-1}\{L(s)/s\}\}}$$

Errors-in-Variables Framework

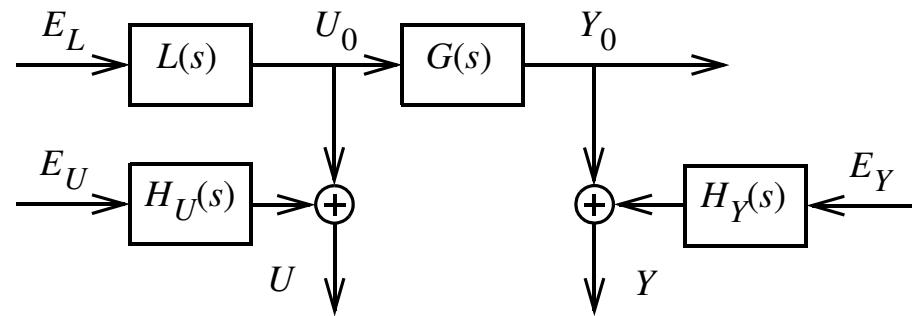


$$G_1(z^{-1}) = \frac{Z\{y_0(nT_s)\}}{Z\{u_0(nT_s)\}} = \frac{\frac{Z\{y_0(nT_s)\}}{Z\{e_L(nT_s)\}}}{\frac{Z\{u_0(nT_s)\}}{Z\{e_L(nT_s)\}}} = \frac{(1 - z^{-1})Z\{L^{-1}\{L(s)G(s)/s\}\}}{(1 - z^{-1})Z\{L^{-1}\{L(s)/s\}\}}$$

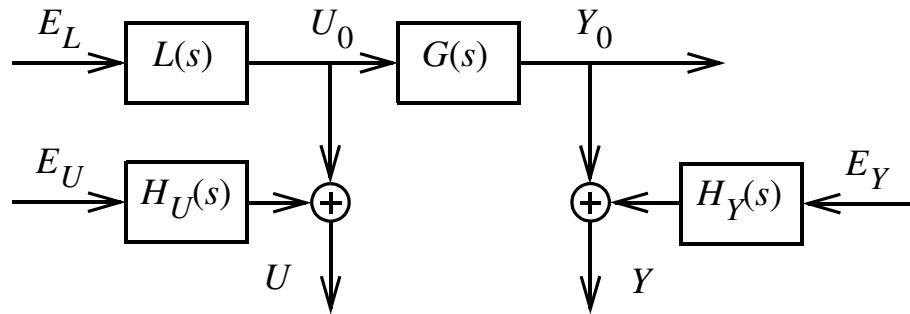
Conclusion

- Exact DT model depends on $L(s)$
- Natural choice = CT modelling

Identifiability Conditions from Second Order Moments

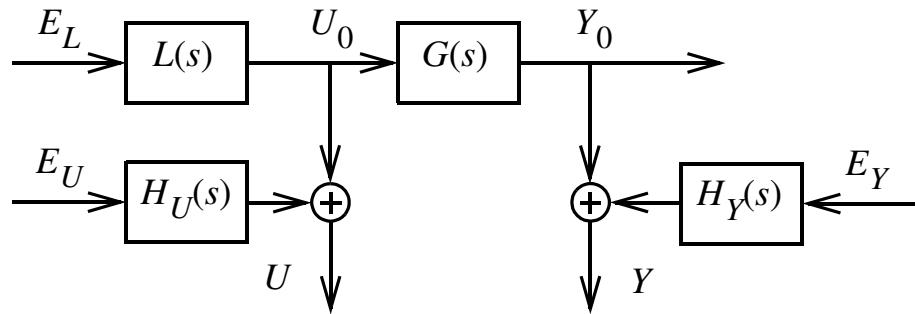


Identifiability Conditions from Second Order Moments



1. $G = \frac{B}{A}$, $L = \frac{P}{Q}$, $H_U = \frac{C_U}{D_U}$, and $H_Y = \frac{C_Y}{D_Y}$ cannot be simplified
2. Monic parametrisation A , P , Q , C_U , D_U , C_Y , and D_Y

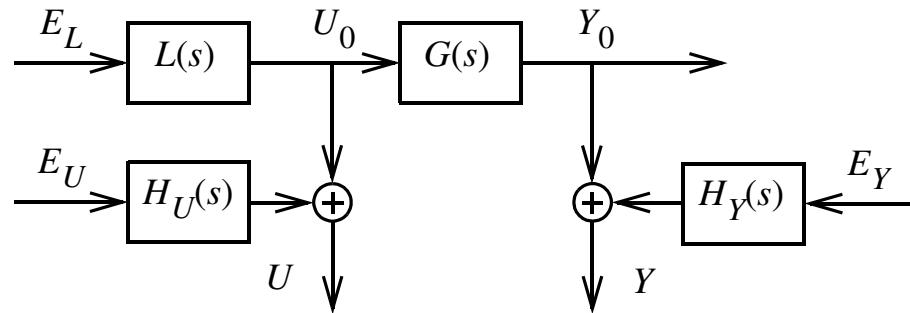
Identifiability Conditions from Second Order Moments



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2. Monic parametrisation A , P , Q , C_U , D_U , C_Y , and D_Y

Not related to EIV

Identifiability Conditions from Second Order Moments

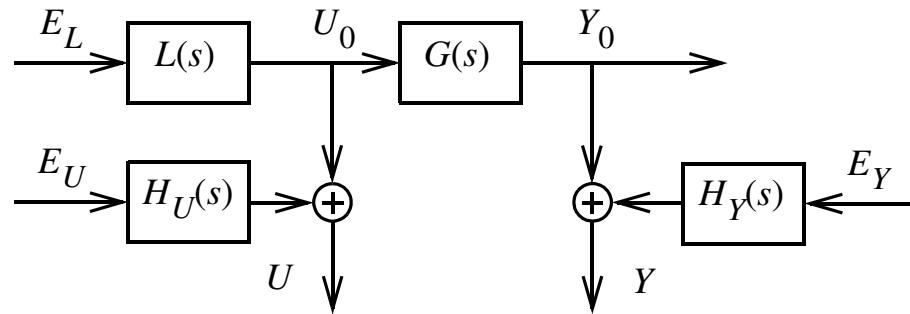


3. $G(s)$ has no quadrant symmetric poles nor zeroes

$$G(s_0) = \begin{cases} 0 \\ \infty \end{cases} \Rightarrow G(-s_0) \neq \begin{cases} 0 \\ \infty \end{cases}$$

4. No pole nor zero of $G(s)$ is respectively a zero or pole of $L(s)L(-s)$

Identifiability Conditions from Second Order Moments



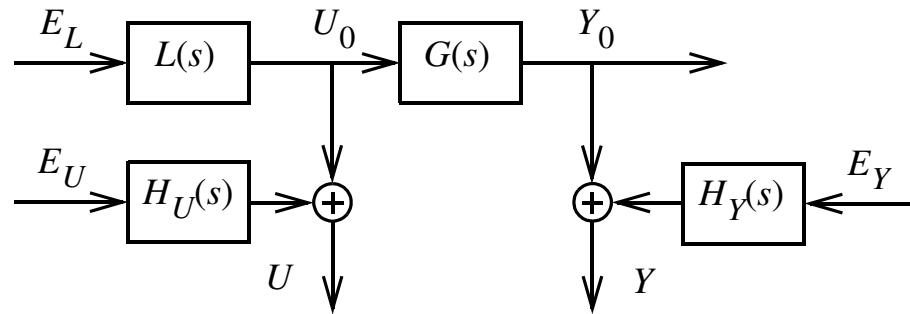
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Finite number of solutions with different model structure

Identifiability Conditions from Second Order Moments

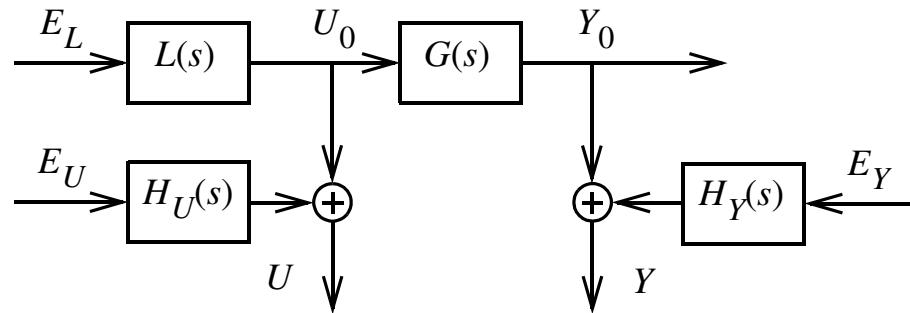


5. One of the following conditions is fulfilled

$$\lim_{s \rightarrow s_0} \frac{H_U(s)H_U(-s)}{L(s)L(-s)} = 0 \text{ for } s_0 = \infty \text{ or pole of } L(s)L(-s)$$

$$\lim_{s \rightarrow s_0} \frac{H_Y(s)H_Y(-s)}{G(s)L(s)G(-s)L(-s)} = 0 \text{ for } s_0 = \infty \text{ or pole of } G(s)L(s)G(-s)L(-s)$$

Identifiability Conditions from Second Order Moments



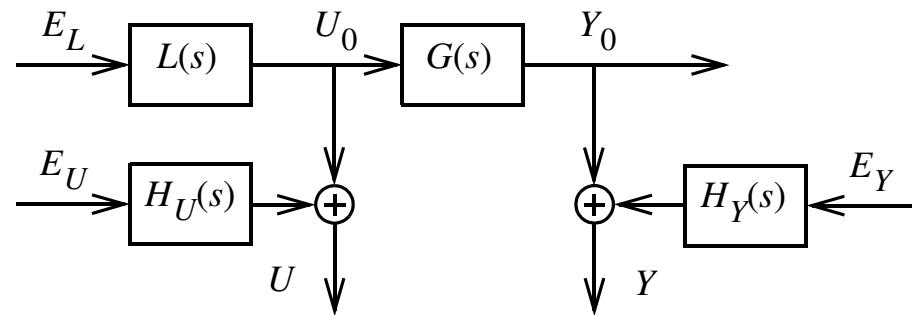
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$$\lim_{s \rightarrow s_0} \frac{H_U(s)H_U(-s)}{L(s)L(-s)} = 0 \text{ for } s_0 = \infty \text{ or pole of } L(s)L(-s)$$

$$\lim_{s \rightarrow s_0} \frac{H_Y(s)H_Y(-s)}{G(s)L(s)G(-s)L(-s)} = 0 \text{ for } s_0 = \infty \text{ or pole of } G(s)L(s)G(-s)L(-s)$$

Infinite number of solutions depending on λ_L

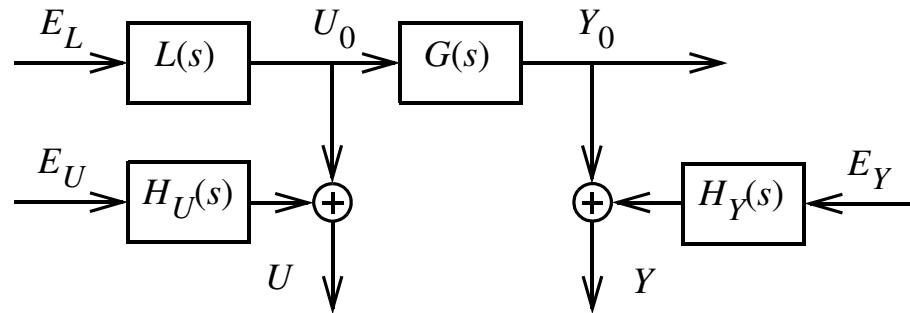
Identifiability Conditions from Second Order Moments



Special cases

1. $L = 1, H_U = 1$, and $H_Y = 1 \Rightarrow$ identifiable iff $G(s)$ is dynamic

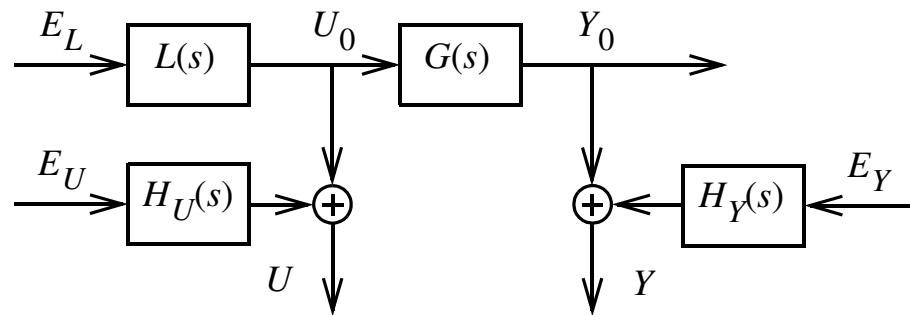
Identifiability Conditions from Second Order Moments



Special cases

1. $L = 1, H_U = 1$, and $H_Y = 1 \Rightarrow$ identifiable iff $G(s)$ is dynamic
2. $G(s)$ is static \Rightarrow identifiable if either $L(s), H_U(s)$, or $H_Y(s)$ depends on s

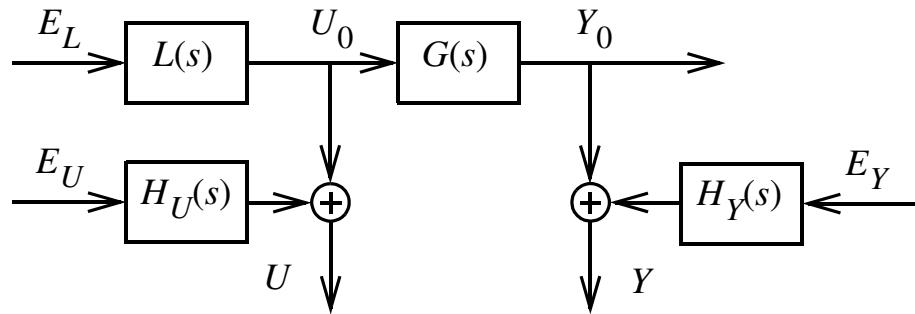
Identifiability Conditions from Second Order Moments



Special cases

1. $L = 1$, $H_U = 1$, and $H_Y = 1 \Rightarrow$ identifiable iff $G(s)$ is dynamic
2. $G(s)$ is static \Rightarrow identifiable if either $L(s)$, $H_U(s)$, or $H_Y(s)$ depends on s
3. $H_Y(s)$ may have the same poles as $G(s) \Rightarrow$ ARX and ARMAX

Frequency Domain Gaussian Maximum Likelihood Estimator



Time domain \rightarrow frequency domain

$$Y(k) = \frac{1}{\sqrt{N}} \sum_{t=0}^{N-1} y(t) e^{-2\pi j k t / N}$$

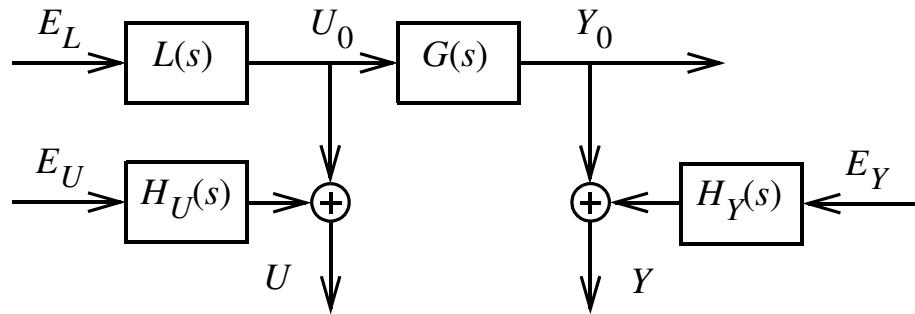
$$U(k) = \frac{1}{\sqrt{N}} \sum_{t=0}^{N-1} u(t) e^{-2\pi j k t / N}$$

Define $Z(k) = [Y(k) \ U(k)]^T$, and $\Lambda = [\lambda_L, \lambda_U, \lambda_Y]^T$

$$\mathbb{E}\{Z(k)|\theta, \Lambda\} = 0$$

$$\text{Cov}(Z(k)|\theta, \Lambda) = C_{Z(k)}(\theta)$$

Frequency Domain Gaussian Maximum Likelihood Estimator



Negative Gaussian log-likelihood

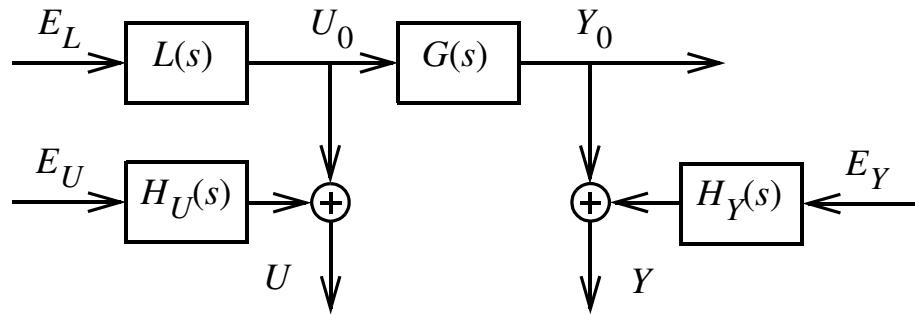
$$V(\theta, \Lambda, Z) = \sum_{k \in \mathbb{K}} \log \det(C_{Z(k)}(\theta)) + \sum_{k \in \mathbb{K}} Z^H(k) C_{Z(k)}^{-1}(\theta) Z(k)$$

with

$$\det(C_{Z(k)}(\theta)) = (|H_Y|^2 \lambda_Y + |GH_U|^2 \lambda_U) |L|^2 \lambda_L + |H_U H_Y|^2 \lambda_U \lambda_Y$$

$$Z^H(k) C_{Z(k)}^{-1}(\theta) Z(k) = \frac{|Y - GU|^2 |L|^2 \lambda_L + |Y|^2 |H_U|^2 \lambda_U + |U|^2 |H_Y|^2 \lambda_Y}{\det(C_{Z(k)}(\theta))}$$

Frequency Domain Gaussian Maximum Likelihood Estimator



Negative Gaussian log-likelihood

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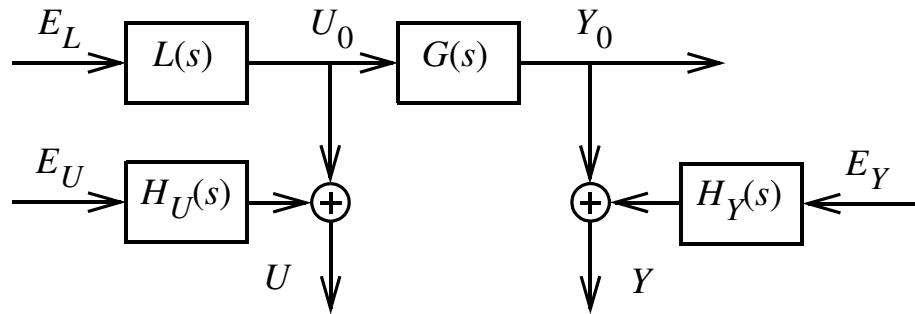
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Discussion

- Spectral factorisation = $\sqrt{\det(C_{Z(k)}(\theta))}$

Frequency Domain Gaussian Maximum Likelihood Estimator



Negative Gaussian log-likelihood

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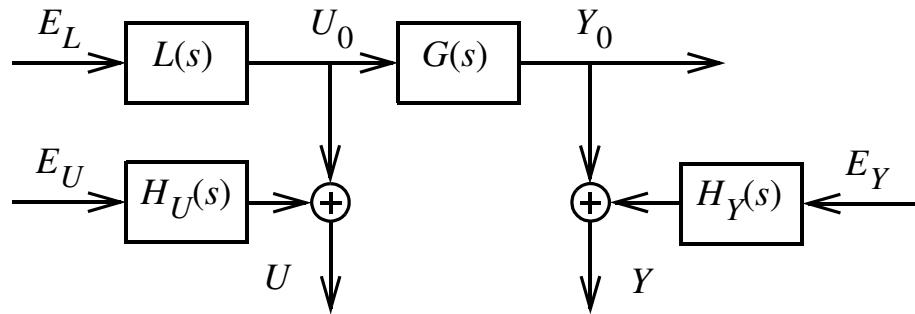
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Discussion

- Spectral factorisation = $\sqrt{\det(C_{Z(k)}(\theta))}$
- Numerical stable Gauss-Newton scheme

Frequency Domain Gaussian Maximum Likelihood Estimator



Negative Gaussian log-likelihood

$$V(\theta, \Lambda, Z) = \sum_{k \in \mathbb{K}} \log \det(C_{Z(k)}(\theta)) + \sum_{k \in \mathbb{K}} Z^H(k) C_{Z(k)}^{-1}(\theta) Z(k)$$

with

$$\det(C_{Z(k)}(\theta)) = (|H_Y|^2 \lambda_Y + |GH_U|^2 \lambda_U) |L|^2 \lambda_L + |H_U H_Y|^2 \lambda_U \lambda_Y$$

$$Z^H(k) C_{Z(k)}^{-1}(\theta) Z(k) = \frac{|Y - GU|^2 |L|^2 \lambda_L + |Y|^2 |H_U|^2 \lambda_U + |U|^2 |H_Y|^2 \lambda_Y}{\det(C_{Z(k)}(\theta))}$$

Discussion

- Spectral factorisation = $\sqrt{\det(C_{Z(k)}(\theta))}$
- Numerical stable Gauss-Newton scheme
- **Exact filtering**

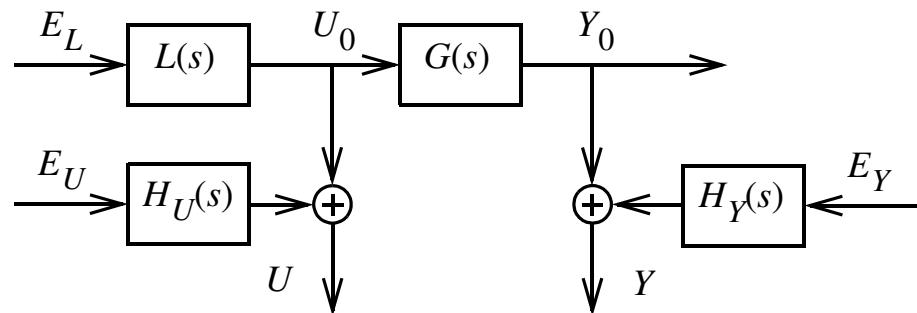
Continuous-Time Simulation Example

$$\lambda_L = 1$$

$$\lambda_U = (0.2)^2$$

$$\lambda_Y = (0.01)^2$$

$$N = 20480$$



$G(s)$ 5th order

$L(s)$ 1st order

$H_U = H_Y = 1$

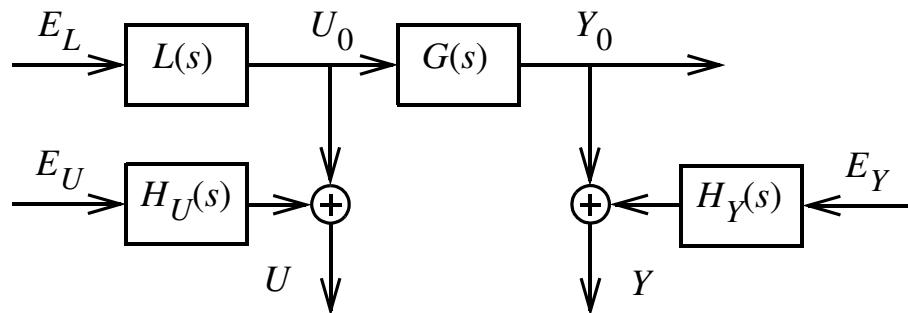
Continuous-Time Simulation Example

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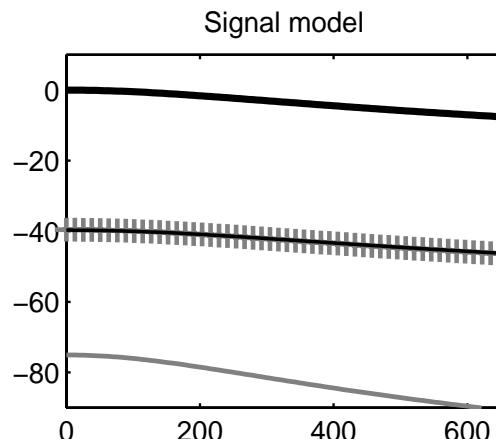
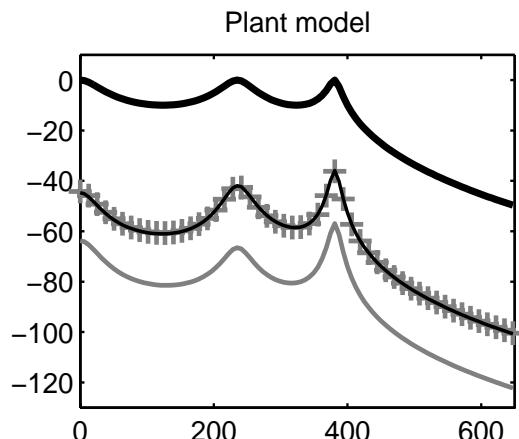
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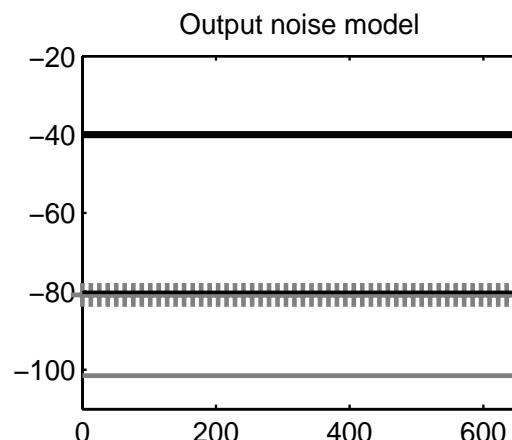
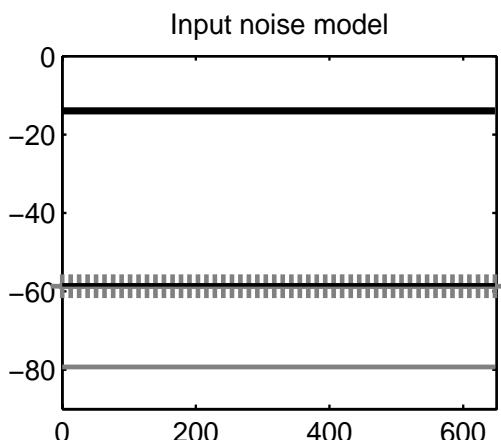
$G(s)$ 5th order

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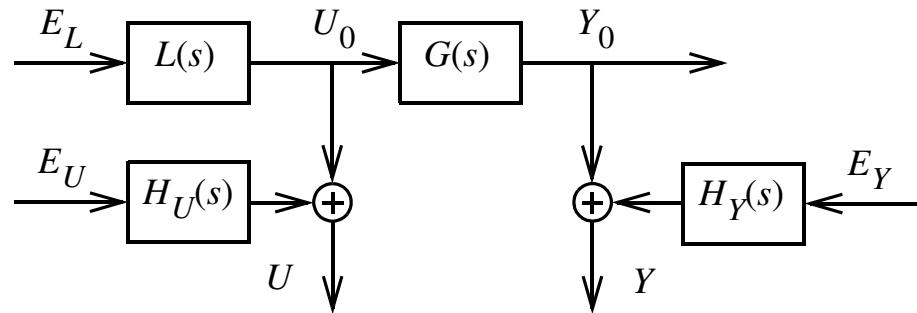
$H_U = H_Y = 1$



- true model
- CR-bound
- complex diff.
- + std ML estim.

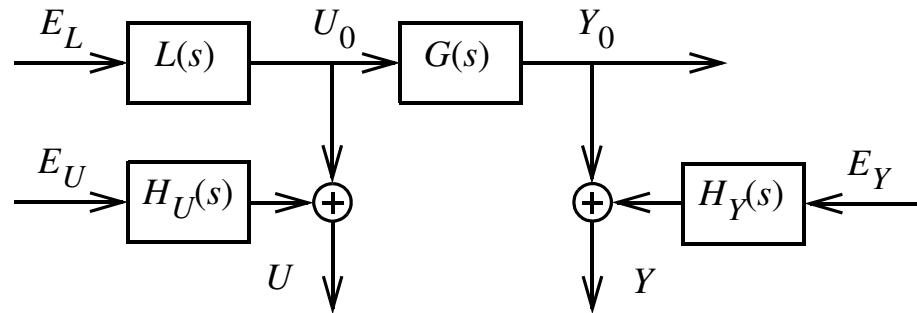


Contributions



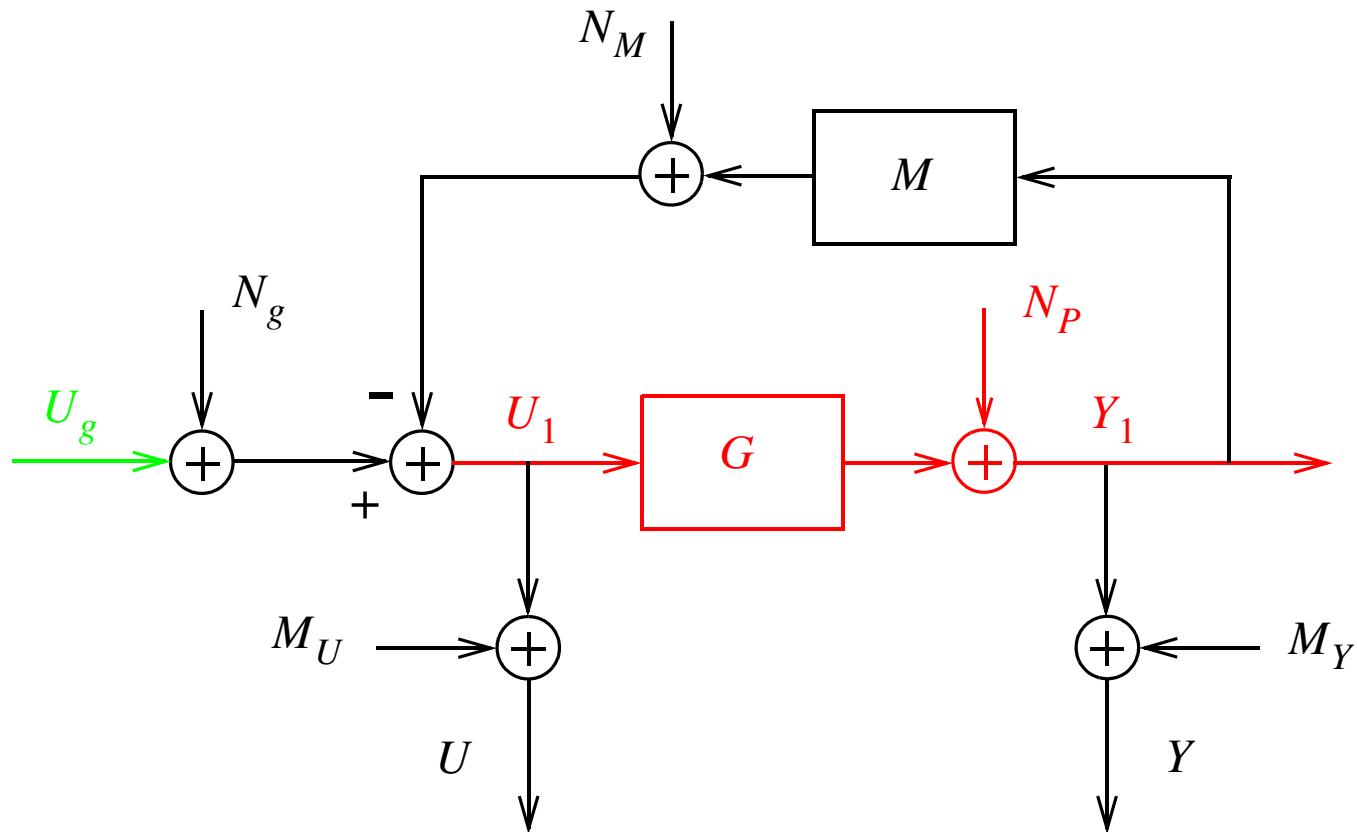
1. Frequency domain Gaussian ML estimator
2. Exact filtering
3. Continuous-time modelling
4. Numerical stable Gauss-Newton minimisation
5. Numerical stable calculation Cramér-Rao lower bound

Open Problems



1. High quality starting values for coloured input/output errors
2. Sensitivity to model errors
3. Validation of the identified models

Concluding Remark



Using periodic excitations

Closed loop EIV is as easy as open loop output error